Synchronous 600 mA Step-Down DC/DC Converter

FEATURES

- Built-in transistors
- **Operating Input Voltage Range**: 2.0 V ~ 6.0 V (A/B/C types) or 1.8 V ~ 6.0 V (D/E/F/G types)
- Output Voltage Range Externally Set: 0.8 V ~ 4.0 V (internally set) or 0.9 ~ 6.0 V (externally set)
- o Output Current: 600 mA
- High Efficiency: 92%
- o Oscillation Frequency: 1.2 MHz, 3 MHz
- Maximum Duty Cycle: 100%
- Operating Modes: PWM, PWM/PFM auto select or PWM/PFM manual select
- Functions: Build-in Current Limit, Load Capacitor Discharge, High Speed Soft start
- Operating Ambient temperature: $-40 \sim +85^{\circ}C$
- **Packages:** SOT-25, USP-6C, USP-6EL, WLP-5-03
- EU RoHS Compliant, Pb Free

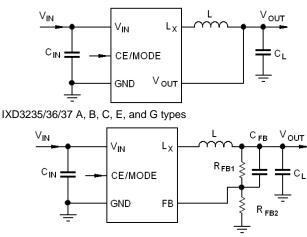
APPLICATION

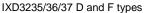
- o Mobile Phones
- o Bluetooth headsets
- o Digital home appliances
- o Office automation equipment
- Various portable equipment

DESCRIPTION

The IXD3235/36/37 series is a group of synchronousrectification type DC/DC converters with a built-in 0.52 Ω N-channel synchronous rectification transistor and 0.42 Ω P-channel switching transistor providing up to 600 mA output current.

TYPICAL APPLICATION CIRCUITS





Operating voltage range is from 2.0 V to 6.0 V (A \sim C types) or 1.8 V to 6.0 V (D \sim G types). For the D/F types, which have a reference voltage of 0.8 V with ± 2.0% accuracy, the output voltage can be set from 0.9 V by using two external resistors.

The A/B/C/E/G types have a fixed output voltage from 0.8 V to 4.0 V in increments of 0.05 V with $\pm 2.0\%$ accuracy. The device requires only an inductor and two externally connected ceramic capacitors. The built-in oscillator, either 1.2 MHz or 3.0 MHz, can be selected.

The IXD3235 operates in PWM mode, the IXD3236 automatically switches between PWM/PFM modes, and the IXD3237 allows switch manually between the PWM and the automatic PWM/PFM switching control modes. This allows fast response, low ripple, and high efficiency over the full range of loads from light to heavy.

The soft start and current control functions are internally optimized. All circuits are disabled in a standby mode to reduce current consumption to less than 1.0 μ A.

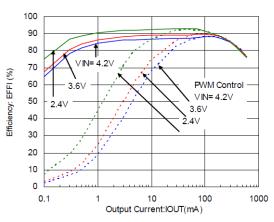
The B/F/G types have a 0.25 ms high-speed softstart for quick turn-on. The built-in Under Voltage Lockout (UVLO) function forces the internal P-channel transistor OFF, when input voltage becomes 1.4 V or lower.

The B to G types have the output capacitor C_L discharge circuitry, which allows fast C_L discharge when IC goes into standby mode.

Device is available in four types of packages: SOT-25, USP-6C, USP-6EL, and WLP-5-03.

TYPICAL PERFORMANCE CHARACTERISTIC

Efficiency vs. Output Current (f_{OSC} = 1.2 MHz, V_{OUT} = 1.8 V) PWM/PFM Automatic Switching mode



ABSOLUTE MAXIMUM RATINGS

PARAMETER		SYMBOL	RATINGS	UNITS
V _{IN} Pin Voltage		V _{IN}	- 0.3 ~ 6.5	V
L _x Pin Voltage		V _{LX}	$-0.3 \sim V_{IN} + 0.3^{1}$	V
V _{OUT} Pin Voltage		V _{OUT}	- 0.3 ~ 6.5	V
FB Pin Voltage		V _{FB}	- 0.3 ~ 6.5	V
CE/MODE Pin Volta	age	V _{CE}	- 0.3 ~ 6.5	V
Lx Pin Current		I _{LX}	±1500	mA
	SOT-25		250	
Dower Dissinction	USP-6C		120	mW
Power Dissipation	USP-6EL		120	IIIVV
	WLP-5-03		750	
Operating Tempera	ture Range	T _{OPR}	- 40 ~ + 85	Do
Storage Temperature Range		T _{STG}	- 50 ~ +125	Do

ELECTRICAL OPERATING CHARACTERISTICS

IXD3235/36/37 A series, V_{OUT} = 1.8 V, Ta = 25°C

PARAMETER		SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
Operating Voltage Range		V _{IN}		2.0	-	6.0	V	0
Output Voltag	e	V _{OUT}	$V_{IN} = V_{CE} = 5.0 \text{ V}, I_{OUT} = 30 \text{ mA}$	1.764	1.800	1.836	V	
Maximum Ou	tput Current	I _{OUT_MAX}	$V_{IN} = V_{OUT(E)} + 2.0 \text{ V}, V_{CE} = 1.0 \text{ V}^{9)}$	600			mA	0
UVLO Voltage	Э	V _{UVLO}	$V_{CE} = V_{IN}, V_{OUT} = 0^{1), 11}$	1.00	1.40	1.78	V	4
Supply	IXD323xA18Cxx				15	33		0
Current	IXD323xA18Dxx	lα	$V_{\text{IN}} = V_{\text{CE}} = 5.0 \text{ V}, V_{\text{OUT}} = V_{\text{OUT(E)}} \text{ x } 1.1 \text{ V}$		21	35	μA	
Standby Curr	ent	I _{STB}	$V_{IN} = 5.0 \text{ V}, \text{ V}_{CE} = 0 \text{ V}, \text{ V}_{OUT} = V_{OUT(E)} \text{ x } 1.1 \text{ V}$		0	1.0	μA	3
Oscillation	IXD323xA18Cxx			1020	1220	1380	kHz	0
Frequency	IXD323xA18Dxx	f _{osc}	$V_{IN} = V_{OUT(E)} + 2 V, V_{CE} = 1.0 V, I_{OUT} = 100 mA$	2550	3000	3460	KHZ	
PFM	IXD323xA18Cxx	. 12)		120	160	200		
Switching Current	IXD323xA18Dxx	$I_{PFM}^{(12)}$ $V_{IN} = V_{CE} = V_{OUT(E)} + 2 V$, , $I_{OUT} = 1 \text{ mA}$ (see table A)	170	220	270	mA		
	I time maximum	t _{PON_MAX} ¹²⁾	$V_{IN} = V_{CE}$ = (see table B), I_{OUT} = 1 mA		2D _{max}	3D _{MAX}		
Maximum Dut	ty Cycle Ratio	D _{MAX}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \times 0.9 \text{ V}$	100			%	2
Minimum Dut	y Cycle Ratio	D _{MIN}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \text{ x } 1.1 \text{ V}$			0	%	2
					92		%	8
Efficiency 2)	IXD323xA18Dxx	EFFI	$V_{IN} = V_{CE} = V_{OUT(E)} + 1.2 \text{ V}, \ I_{OUT} = 100 \text{ mA}$		86		%	
L _x "H" ON Re	sistance 1 ³⁾	R _{LXH1}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.35	0.55	Ω	4
L _x "H" ON Re	sistance 2 ³⁾	R _{LXH2}	$V_{IN} = V_{CE} = 3.6 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.42	0.67	Ω	4
L _x "L" ON Res	sistance 1 ⁴⁾	R _{LXL1}	$V_{IN} = V_{CE} = 5.0 \text{ V}$		0.45	0.65	Ω	4
L _x "L" ON Res	sistance 24)	R _{LXL2}	$V_{IN} = V_{CE} = 3.6 \text{ V}$		0.52	0.77	Ω	4
L _x "H" Leakag	je Current ⁵⁾	I _{LXH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	6
L _x "L" Leakag	e Current ⁵⁾	I _{LXH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	6
Current Limit ¹	0)	I _{LIM}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \times 0.9 \text{ V}^{8)}$	900	1050	1350	mA	
	e Temperature	ΔV_{OUT}	$-40^{\circ}C \le T_{OPR} \le 85^{\circ}C, I_{OUT} = 30 \text{ mA}$		±100		ppm/ºC	2
Characteristic		$V_{OUT} * \Delta T_{OPR}$ V_{CEH}	V _{OUT} = 0 V	0.65		6.0	v	(4)
CE "L" Voltage ¹⁵⁾		V _{CEH}	V _{OUT} = 0 V	0.65		0.0	V	(4) (4)
0	e start Voltage ^{6), 13)}	V _{CEL}	$V_{OUT} = 0.V$ $I_{OUT} = 1 \text{ mA}$	U		0.25 V _{IN} -1.0	v	U
PWM/PFM m				V _{IN} –		v _{IN} -1.0		
Start Voltage		V _{PFM}	I _{OUT} = 1 mA	0.25				
CE "H" Currer	nt	I _{ENH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}$	-0.1		0.1	μA	6



CE "L" Curr	rent	I _{ENL}	$V_{IN} = 5.0 \text{ V}, V_{CE} = 0 \text{ V}, V_{OUT} = 0 \text{ V}$	-0.1		0.1	μA	6
Soft-Start	IXD323xA18Cxx	+	I _{OUT} = 1 mA (see table C)		1.0	2.5	m 0	2
Time	IXD323xA18Dxx	t _{SS}	$I_{OUT} = I IIIA (see table C)$	0.5	0.9	2.5	ms	Ø
Latch Time	7)	t _{LAT}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0.8 \text{ x} V_{OUT(E)}, L_X \text{ short with}$ 1 Ω resistor to ground	1.0		20.0	ms	
Short Prote Voltage	ction Threshold	V _{SHORT}	$V_{IN} = V_{CE} = 5.0 \text{ V}, L_X \text{ short with 1 } \Omega \text{ resistor to}$ ground	0.675	0.900	1.150	V	

NOTE:

Test conditions: Unless otherwise stated, V_{IN} = 5.0 V, $V_{OUT(E)}$ = Nominal Voltage

- 1) Including hysteresis operating voltage range
- 2) EFFI = {(output voltage × output current) / (input voltage × input current)} × 100%
- 3) ON resistance (Ω) = (V_{IN} Lx pin measurement voltage) / 100mA
- 4) Design target value
- 5) A 10µA (maximum) current may leak at high temperature
- 6) The CE/MODE pin of the IXD3237A series functions also as an external switching pin between PWM and PWM/PFM control. Control is switched to the automatic PWM/PFM switching mode when the CE/MODE pin voltage is equal to or greater than V_{IN} minus 0.3 V, and to the PWM mode when the CE/MODE pin voltage is equal to or lower than VIN minus 1.0V. However, it should be equal to or greater than V_{CEH}
- Time from moment when V_{OUT} is shorted to GND via 1 Ω resistor to the moment, when Current Limit generates pulse stopping L_x oscillations
- 8) When V_{IN} is less than 2.4 V, current limit may not be reached because of voltage drop across ON resistance
- 9) When the difference between input and output voltage is small, some cycles may be skipped completely before current maximizes. If load current increases in this state, output voltage will decrease because of the voltage drop across P-channel transistor
- 10) Current limit denotes the level of an inductor peak current
- 11) Voltage, when L_x pin voltage is "L"=+0.1 V ~ -0.1 V
- 12) Not for IXD3235 series, because they have PWM mode only
- 13) The IXD3237 series only
- 14) Voltage at which L_x pin state changes from "L" to "H" = $V_{IN} \sim V_{IN} 1.2$ V"
- 15) Voltage at which L_x pin state changes from "H" to "L" "=+0.1 V ~ -0.1 V

ELECTRICAL OPERATING CHARACTERISTICS (CONTINUED)

IXD3235/36/37 B/C/E/G series, V_{OUT} = 1.8 V, Ta = 25°C

Р	ARAME	TER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUI
Operating Voltage B/C series				2.0	-	6.0		_	
Range E/G series		V _{IN}		1.8		6.0	V	0	
Output Vo	tage	l	V _{OUT}	$V_{IN} = V_{CE} = 5.0 \text{ V}, I_{OUT} = 30 \text{ mA}$	1.764	1.800	1.836	V	
Maximum	Output C	urrent	I _{OUT_MAX}	$V_{IN} = V_{OUT(E)} + 2.0 \text{ V}, V_{CE} = 1.0 \text{ V}^{9)}$	600			mA	0
UVLO Volt	age		V _{UVLO}	$V_{CE} = V_{IN}, V_{OUT} = V_{OUT(E)} \times 0.5 \text{ V}^{(1), (11), (16)}$	1.00	1.40	1.78	V	4
Supply	IXD	323xx18Cxx	1	V _{IN} = V _{CE} = 5.0 V, V _{OUT} = V _{OUT(E)} x 1.1 V		15	33		0
Current	IXD	323xx18Dxx	Ι _Q	$v_{IN} = v_{CE} = 5.0 v, v_{OUT} = v_{OUT(E)} x 1.1 v$		21	35	μA	
Standby C	urrent		I _{STB}	$V_{\text{IN}} = 5.0 \text{ V}, V_{\text{CE}} = 0 \text{ V}, V_{\text{OUT}} = V_{\text{OUT(E)}} \text{ x 1.1 V}$		0	1.0	μA	3
Oscillation	IXD	323xx18Cxx	£	\/ \/ · · 2\/.\/ 1.0.\/ 100 mA	1020	1220	1380		2
Frequency	IXD	323xx18Dxx	f_{OSC} $V_{IN} = V_{OUT(E)} + 2 V, V_{CE} = 1.0 V, I_{OUT} = 100 mA$	2550	3000	3460	kHz		
PFM	IXD	323xx18Cxx	12)		120	160	200		
Switching Current	IXD	323xx18Dxx	I _{PFM} ¹²⁾	$V_{IN} = V_{CE} = V_{OUT(E)} + 2 V$, , $I_{OUT} = 1 \text{ mA}$ (see table A)	170	220	270	mA	
P-channel	ON time	maximum	t _{PON_MAX} ¹²⁾	$V_{IN} = V_{CE}$ = (see table B), I_{OUT} = 1 mA		2D _{max}	$3D_{\text{MAX}}$		
Maximum	Duty Cyc	le Ratio	D _{MAX}	$V_{\text{IN}} = V_{\text{CE}} = 5.0 \text{ V}, V_{\text{OUT}} = V_{\text{OUT(E)}} \text{ x } 0.9 \text{ V}$	100			%	2
Minimum [Duty Cyc	e Ratio	D _{MIN}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \text{ x 1.1 V}$			0	%	2
Efficiency	2) IXD	323xx18Cxx				92		0/	8
Emclency	IXD	323xx18Dxx	EFFI	$V_{IN} = V_{CE} = V_{OUT(E)} + 1.2 \text{ V}, I_{OUT} = 100 \text{ mA}$		86		%	
L _x "H" ON Resistance 1 ³⁾		R _{LXH1}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.35	0.55	Ω	4	
L _x "H" ON Resistance 2 ³⁾		R _{LXH2}	$V_{IN} = V_{CE} = 3.6 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.42	0.67	Ω	4	
L _x "L" ON Resistance 1 ⁴⁾		R _{LXL1}	$V_{IN} = V_{CE} = 5.0 \text{ V}$		0.45	0.65	Ω	4	
L _X "L" ON	Resistan	ce 2 ⁴⁾	R _{LXL2}	$V_{IN} = V_{CE} = 3.6 \text{ V}$		0.52	0.77	Ω	4
L _x "H" Lea	kage Cu	rent ⁵⁾	I _{LXH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	6
L _x "L" Leal	age Cur	rent ⁵⁾	I _{LXH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	6
Current Lir	nit ¹⁰⁾		I _{LIM}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \text{ x } 0.9 \text{ V}^{8)}$	900	1050	1350	mA	
Output Vo		nperature	ΔV_{OUT}	$-40^{\circ}C \le T_{OPR} \le 85^{\circ}C, I_{OUT} = 30 \text{ mA}$		±100		ppm/ºC	2
Characteri			$V_{OUT} * \Delta T_{OPR}$		0.05		0.0		(4)
CE "H" Vo			V _{CEH}	V _{OUT} = 0 V	0.65		6.0	V	
CE "L" Vol	0	6), 13)	V _{CEL}	V _{OUT} = 0 V	0		0.25	V	4
PWM mod PWM/PFM		oltage	V _{PWM}	I _{OUT} = 1 mA	V _{IN} –		V _{IN} -1.0		
Start Volta			V _{PFM}	$I_{OUT} = 1 \text{ mA}$	0.25				
CE "H" Cu			I _{ENH}	$V_{\text{IN}} = V_{\text{CE}} = 5.0 \text{ V}, V_{\text{OUT}} = 0 \text{ V}$	-0.1		0.1	μA	6
CE "L" Cu			I _{ENL}	$V_{IN} = 5.0 \text{ V}, V_{CE} = 0 \text{ V}, V_{OUT} = 0 \text{ V}$	-0.1		0.1	μA	6
	IXD323	xB(G)18Cxx				0.25	0.4		
Soft-Start		xC(E)18Cxx	· ·	1 = 1 = 1 = 1	0.5	1.0	2.5		
Time	IXD323	xB(G)18Dxx	t _{ss}	I _{OUT} = 1 mA (see table C)		0.32	0.50	ms	2
IXD323xC(E)18Dxx			0.5	0.9	2.5				
Latch Time	e ⁷⁾		t _{LAT}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0.8 \text{ x} V_{OUT(E),} L_x \text{ short with}$ 1 Ω resistor to ground	1.0		20.0	ms	
Short Prot	ection	B/C series	V	$V_{IN} = V_{CE} = 5.0 \text{ V}, L_x \text{ short with 1 } \Omega \text{ resistor to}$	0.675	0.900	1.150	17	
Threshold		E/G series		0.338	0.450	0.563	V		
C _L Dischar	ge Resis	tance	R _{DCL}	$V_{IN} = V_{LX} = 5.0 \text{ V}, V_{CE} = 0 \text{ V}, V_{OUT} \text{ - open}$	200	300	450	Ω	

NOTE:

Test conditions: Unless otherwise stated, V_{IN} = 5.0 V, $V_{OUT(E)}$ = Nominal Voltage

Including hysteresis operating voltage range 1)

- 2) 3) 4)́

Design target value



5) A 10µA (maximum) current may leak at high temperature

6) The CE/MODE pin of the IXD3237A series functions also as an external switching pin between PWM and PWM/PFM control. Control is switched to the automatic PWM/PFM switching mode when the CE/MODE pin voltage is equal to or greater than V_{IN} minus 0.3 V, and to the PWM mode when the CE/MODE pin voltage is equal to or lower than VIN minus 1.0V. However, it should be equal to or greater than V_{CEH}

- Time from moment when V_{OUT} is shorted to GND via 1 Ω resistor to the moment, when Current Limit generates pulse stopping L_x oscillations
- 8) When V_{IN} is less than 2.4 V, current limit may not be reached because of voltage drop across ON resistance
- 9) When the difference between input and output voltage is small, some cycles may be skipped completely before current maximizes. If load current increases in this state, output voltage will decrease because of the voltage drop across P-channel transistor
 40) Current limit denotes the lower of the voltage across P-channel transistor
- 10) Current limit denotes the level of an inductor peak current
- 11) Voltage, when L_x pin voltage is "L"=+0.1 V ~ -0.1 V
- 12) Not for IXD3235 series, because they have PWM mode only
- 13) The IXD3237 series only
- 14) Voltage at which L_x pin state changes from "L" to "H" = $V_{IN} \sim V_{IN} 1.2$ V"
- 15) Voltage at which L_x pin state changes from "H" to "L" "=+0.1 V ~ -0.1 V
- 16) Voltage at which V_{OUT} becomes more than V_{IN} , while V_{IN} is rising from 0 V to $V_{OUT (E)} \times 0.5 V$



ELECTRICAL OPERATING CHARACTERISTICS (CONTINUED)

IXD3235/36/37 D/F series, V_{OUT} = 1.8 V, Ta = 25°C

PA	RAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
Operating Voltage Range		V _{IN}		1.8	-	6.0	V	0
FB Voltage		V _{FB}	$V_{IN} = V_{CE} = 5.0 \text{ V}, I_{OUT} = 30 \text{ mA}$	1.784	1.800	1.816	V	
Maximum C	utput Current	I _{OUT_MAX}	$V_{IN} = V_{OUT(E)} + 2.0 \text{ V}, V_{CE} = 1.0 \text{ V}^{9)}$				mA	0
UVLO Volta	ge	V _{UVLO}	$V_{CE} = V_{IN}, V_{OUT} = 0^{1), 11}$	1.00	1.40	1.78	V	4
Supply	IXD323xx18Cxx				15			0
Current	IXD323xx18Dxx	- Ι _Q	$V_{\text{IN}} = V_{\text{CE}} = 5.0 \text{ V}, V_{\text{OUT}} = V_{\text{OUT(E)}} \times 1.1 \text{ V}$		21	35	μA	
Standby Cu	rrent	I _{STB}	$V_{IN} = 5.0 \text{ V}, V_{CE} = 0 \text{ V}, V_{OUT} = V_{OUT(E)} \text{ x } 1.1 \text{ V}$		0	1.0	μA	3
Oscillation	IXD323xx18Cxx	4		1020	1220	1380		2
Frequency	IXD323xx18Dxx	f _{osc}	$V_{IN} = V_{OUT(E)}$ + 2 V, V_{CE} = 1.0 V, I_{OUT} = 100 mA	2550	3000	3460	kHz	
PFM	IXD323xx18Cxx	. 12)		120	160	200		
Switching Current	IXD323xx18Dxx	$I_{PFM}^{(12)}$ $V_{IN} = V_{CE} = V_{OUT(E)} + 2 V_{, I}$, $I_{OUT} = 1 \text{ mA}$ (see table A)	$V_{IN} = V_{CE} = V_{OUT(E)} + 2 V$, , $I_{OUT} = 1 \text{ mA}$ (see table A)	170	220	270	mA	
	N time maximum	t _{PON_MAX} ¹²⁾	$V_{IN} = V_{CE}$ = (see table B), I_{OUT} = 1 mA		2D _{max}	3D _{MAX}		
Maximum D	uty Cycle Ratio	D _{MAX}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \times 0.9 \text{ V}$	100			%	2
Minimum D	uty Cycle Ratio	D _{MIN}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \times 1.1 \text{ V}$			0	%	2
	IXD323xx18Cxx				92			8
Efficiency 2)	IXD323xx18Dxx	EFFI	$V_{IN} = V_{CE} = V_{OUT(E)} + 1.2 \text{ V}, \ I_{OUT} = 100 \text{ mA}$		86		%	
L _X "H" ON F	esistance 1 ³⁾	R _{LXH1}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.35	0.55	Ω	4
L _x "H" ON Resistance 2 ³⁾		R _{LXH2}	$V_{IN} = V_{CE} = 3.6 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.42	0.67	Ω	4
L _x "L" ON Resistance 1 ⁴⁾		R _{LXL1}	$V_{IN} = V_{CE} = 5.0 \text{ V}$		0.45	0.65	Ω	4
L _X "L" ON R	esistance 24)	R _{LXL2}	$V_{IN} = V_{CE} = 3.6 \text{ V}$		0.52	0.77	Ω	4
L _X "H" Leak	age Current ⁵⁾	I _{LXH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	6
L _x "L" Leaka	age Current ⁵⁾	I _{LXH}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	6
Current Lim	it ¹⁰⁾	I _{LIM}	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \times 0.9 \text{ V}^{8)}$	900	1050	1350	mA	
	age Temperature	ΔV_{OUT}	$-40^{\circ}C \le T_{OPR} \le 85^{\circ}C, I_{OUT} = 30 \text{ mA}$		±100		ppm/ºC	2
Characteris CE "H" Volt		$V_{OUT} * \Delta T_{OPR}$		0.65		6.0	V	4
CE H VOIL	-	V _{CEH}	V _{OUT} = 0 V	0.65 0		6.0	V	4
	Start Voltage ^{6), 13)}	V _{CEL}		0		0.25	v	æ
PWM/PFM		V _{PWM}	I _{OUT} = 1 mA	V _{IN} –		V _{IN} -1.0		
Start Voltag		V _{PFM}	I _{OUT} = 1 mA	0.25				
CE "H" Curi	ent	I _{ENH}	$V_{\text{IN}} = V_{\text{CE}} = 5.0 \text{ V}, V_{\text{OUT}} = 0 \text{ V}$	-0.1		0.1	μA	6
CE "L" Curr	ent	I _{ENL}	$V_{\text{IN}} = 5.0 \text{ V}, V_{\text{CE}} = 0 \text{ V}, V_{\text{OUT}} = 0 \text{ V}$	-0.1		0.1	μA	6
	IXD323xD18Cxx			0.5	1.0	2.5		
Soft-Start			$I_{OUT} = 1 \text{ mA}$ (see table C)		0.25	0.40	me	0
Time	IXD323xD18Dxx	t _{ss}	IOUI - I THA (SEE LADIE C)	0.5	1.0	2.5	ms	é
	IXD323xF18Dxx				0.25	0.40		
Latch Time		t _{LAT}	V_{IN} = V_{CE} = 5.0 V, V_{OUT} = 0.8 x $V_{\text{OUT}(\text{E})}$ Lx short with 1 Ω resistor to ground	1.0		20.0	ms	
Voltage	ction Threshold	V _{SHORT}	$V_{IN} = V_{CE} = 5.0 \text{ V}, L_X \text{ short with 1 } \Omega \text{ resistor to}$ ground	0.675	0.900	1.150	V	
C _L Discharg	e Resistance	R _{DCL}	$V_{IN} = V_{LX} = 5.0 \text{ V}, V_{CE} = 0 \text{ V}, V_{OUT}$ - open	200	300	450	Ω	

NOTE:

Test conditions: Unless otherwise stated, $V_{IN} = 5.0 \text{ V}$, $V_{OUT(E)} = Nominal Voltage$

Including hysteresis operating voltage range 1)

2) 3) EFFI = {(output voltage × output current) / (input voltage × input current)} × 100%

ON resistance (Ω) = (V_{IN} - Lx pin measurement voltage) / 100mA

Design target value 4)

A 10µA (maximum) current may leak at high temperature 5)

The CE/MODE pin of the IXD3237A series functions also as an external switching pin between PWM and PWM/PFM control. Control is switched to the automatic PWM/PFM switching mode when the CE/MODE pin voltage is equal to or greater than V_{IN} minus 0.3 V, 6)



and to the PWM mode when the CE/MODE pin voltage is equal to or lower than VIN minus 1.0V. However, it should be equal to or greater than V_{CEH}

- Time from moment when V_{OUT} is shorted to GND via 1 Ω resistor to the moment, when Current Limit generates pulse stopping L_x oscillations
- 8) When V_{IN} is less than 2.4 V, current limit may not be reached because of voltage drop across ON resistance
- 9) When the difference between input and output voltage is small, some cycles may be skipped completely before current maximizes. If load current increases in this state, output voltage will decrease because of the voltage drop across P-channel transistor
 40) Current limit denotes the lowel of an industry pack surrent.
- 10) Current limit denotes the level of an inductor peak current 11) Voltage, when L_X pin voltage is "L"=+0.1 V ~ -0.1 V
- 12) Not for IXD3235 series, because they have PWM mode only
- 13) The IXD3237 series only
- 14) Voltage at which L_x pin state changes from "L" to "H" = $V_{IN} \sim V_{IN} 1.2$ V"
- 15) Voltage at which L_x pin state changes from "H" to "L" "=+0.1 V ~ -0.1 V

TABLE A

PFM Switching Current (I_{PFM}) vs. Oscillation Frequency and Setting Voltage

SETTING VOLTAGE	f _{osc} = 1.2 MHz			f _{osc} = 3.0 MHz			
SETTING VOLTAGE	MIN	ТҮР	MAX	MIN	ТҮР	MAX	
$V_{OUT(E)} \le 1.2 V$	140	180	240	190	260	350	
$1.2 \text{ V} < \text{V}_{\text{OUT(E)}} \le 1.75 \text{ V}$	130	170	220	180	240	300	
V _{OUT(E)} ≥ 1.8 V	120	160	200	170	220	270	

TABLE B

Input Voltage (VIN) for Measuring P-channel ON time maximum tPON MAX

f _{osc}	1,2 MHZ	3 MHZ
V _{IN}	V _{OUT(E)} + 0.5 V	V _{OUT(E)} +1.0 V

NOTE:

Example: When $V_{OUT(E)} = 1.2V$ and $f_{OSC} = 1.2$ MHz, V_{IN} should be 1.7 V, however, V_{IN} should be at least 2.0 V if the minimum operating voltage is 2.0 V

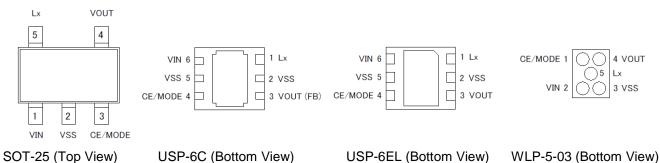
TABLE C

Soft-Start Time vs. Setting Voltage and Oscillation Frequency (IXD3235/36/37 B and G Series only)

SERIES	fosc SETTING VOLTAGE, V		SOFT	ME, μS	
SERIES	f _{osc}	SETTING VOLTAGE, V	MIN	TYP	MAX
		$0.8 \le V_{OUT(E)} < 1.75$		250	400
IXD3235B/G		1.5 ≤ V _{OUT(E)} < 1.8		320	500
IXD3237B/G	1.2 MHz	$1.8 \le V_{OUT(E)} < 2.5$		250	400
		$2.5 \le V_{OUT(E)} < 4.0$		320	500
IXD3236B/G		$0.8 \le V_{OUT(E)} < 2.5$		250	400
1AD3230B/G		$2.5 \le V_{OUT(E)} < 4.0$		320	500
IXD3235/36/37 B/G	3.0 MHz	$0.8 \le V_{OUT(E)} < 1.8$		250	400
IND3233/30/37 D/G		1.8 ≤ V _{OUT(E)} < 4.0		320	500



PIN CONFIGURATION



NOTE:

The dissipation pad for the USP-6C and USP-6EL packages should be soldered in recommended mount pattern and metal masking to enhance mounting strength and heat release. If the pad needs to be connected to other pins, it should be connected to the V_{SS} (No 2 and No 5) pins. V_{SS} pins (No. 2 and 5) should be tied together.

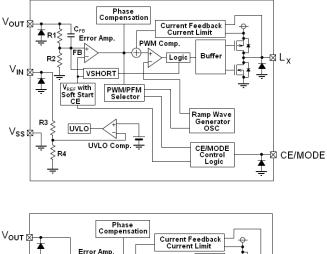
PIN ASSIGNMENT

	PIN NUMBER		PIN NAME	FUNCTIONS
SOT-25	USP-6C/USP-6EL	WLP-5-03		FUNCTIONS
1	6	2	V _{IN}	Power Input
2	2, 5	3	V _{SS}	Ground
3	4	1	CE/MODE	Enable (Active HIGH), Mode Selection Pin
4	3	4	V _{OUT} (FB)	Fixed Output Voltage - A/B/C/E/G series (Output Voltage Sense Pin - D/F series)
5	1	5	Lx	Switching Node

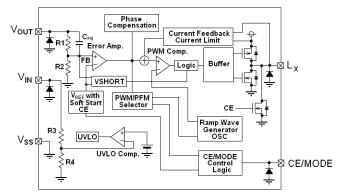


BLOCK DIAGRAMS

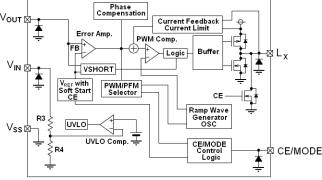
IXD3235/36/37 A Series



IXD3235/36/37 B/C/E/G Series



IXD3235/36/37 D/F Series



Internal diodes include an ESD protection and a parasitic diode

BASIC OPERATION

The IXD3235/36/37 series consists of a Reference Voltage source, Ramp Wave Generator, Error Amplifier, PWM Comparator, Phase Compensation circuit, output voltage resistive divider, P-channel switching transistor, N-channel transistor for the synchronous switch, Current Limiter circuit, UVLO circuit, and others. (See the block diagram above.)

The Error Amplifier compares output voltage divided by internal (external for D/F versions) resistors R_{FB1}/R_{FB2} with the internal reference voltage. Amplified difference between these two signals applies to the one input of the PWM Comparator, while ramp voltage from the Ramp Wave Generator applies to the second input. Resulting PWM pulse determines switching transistor ON time. It goes through the Buffer and it appears at the gate of the internal P-channel switching transistor. This continuous process stabilizes output voltage.

The Current Feedback circuit monitors current of the P-channel transistor at each switching cycle, and modulates output signal from the Error Amplifier to provide additional feedback. This guarantees a stable converter operation even with low ESR ceramic load capacitor.

Reference Voltage Source

The Reference Voltage Source provides the reference voltage to ensure stable output voltage of the DC/DC converter.

Ramp Wave Generator

The Ramp Wave Generator produces ramp waveform signal needed for PWM operation, and signals to synchronize all the internal circuits. It operates at internally fixed 1.2 MHz or 3.0 MHz frequency.

Error Amplifier

The Error Amplifier monitors output voltage through resistive divider connected to V_{OUT} (FB) pin. If output voltage falls below preset value and Error Amplifier's input signal becomes less than internal reference voltage, the Error Amplifier's output signal increases. That results in wider PWM pulse and respectively longer ON time for switching transistor to increase output voltage. The gain and frequency characteristics of the error amplifier output are fixed internally to optimize IC performance.

Current Limiter

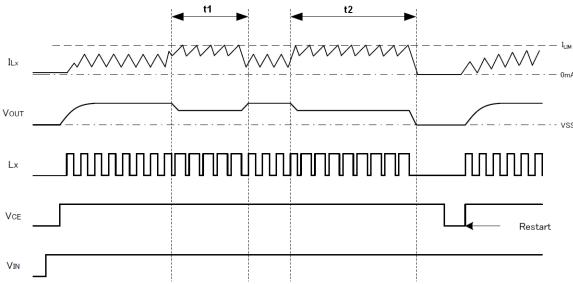
The Current Limiter circuit monitors current flowing through the P-channel transistor connected to the Lx pin, and combines function of the current limit and operation suspension.

When transistor's current is greater than a specified level, the Current Limiter turns off P-channel transistor immediately. After that, the Current Limiter turns off too, returning to monitoring mode.

The driver transistor turns on at the next cycle, but the Current Limiter will turn it off immediately if an over current exists. When the over current state is eliminated, the IC resumes its normal operation.

The IC waits for end of the over current state repeating above steps (t1 on figure below). If an over-current state continues for a few ms with IC repeatedly performing above steps, the Current Limiter latches the P-channel transistor in OFF state, and IC suspends operations (t2 on figure below). To restart IC operation after this condition, either EN pin should be toggled H - L - H, or V_{IN} pin voltage should be set below UVLO to resume operations from soft start.

The suspension mode is not a standby mode. In the suspension mode, pulse output is suspended; however, internal circuitries remain in operation mode consuming power.



Short-Circuit Protection

The short-circuit protection monitors the R_{FB1}/R_{FB2} divider voltage (FB point in the block diagram). If output is accidentally shorted to the ground, FB voltage starts falling. When this voltage becomes less than half of the reference voltage (V_{REF}) and P-channel switching transistor's current is more than the I_{LIM} threshold, the Short-Circuit Protection turns off and latches quickly the P-channel transistor.

At D/E/F/G series, Short Circuit Protection starts once FB voltage becomes less than 0.25 of reference voltage (V_{REF}), disregard to transistor's current.

To restart IC operation after this condition, either EN pin should be toggled H - L - H, or V_{IN} pin voltage should be set below UVLO to resume operations from soft start.

The sharp load transients creating a voltage drop at the V_{OUT} , propagate to the FB point through C_{FB} , that may result in Short Circuit protection operating at voltages higher than 1/2 V_{REF} voltage.

UVLO Circuit

When the V_{IN} pin voltage becomes 1.4V or lower, the P-channel transistor is forced OFF to prevent false pulse output caused by unstable operation of the internal circuitry. When the V_{IN} pin voltage becomes 1.8 V or higher,



switching operations resume with the soft start. The soft start function operates even when the V_{IN} voltage falls below the UVLO threshold for a very short time. The UVLO circuit does not cause a complete shutdown of the IC, but causes pulse output to be suspended; therefore, the internal circuitry remains in operation.

PFM Switch Current

In PFM mode, the IC keeps the P-channel transistor on until inductor current reaches a specified level (I_{PFM}).

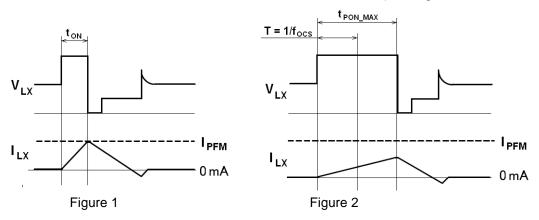
P-channel transistor's ON time is equal

$$t_{ON} = L \times I_{PFM} / (V_{IN} - V_{OUT}), \mu s,$$

where L is an inductance in μ H, and I_{PFM} is a current limit in A.

PFM Duty Limit

In PFM mode, P-channel ON time maximum (t_{PON_MAX}) is set to $2D_{MAX}$, i.e. two periods of the switching frequency. Therefore, under conditions, when the ON time increases (i.e. step-down ratio is small), it is possible that P-channel transistor to be turned off, even when inductor current does not reach to I_{PFM} . (See Figures 1 and 2 below)



C_L High Speed Discharge

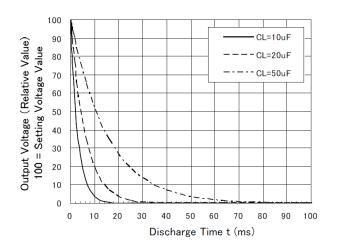
The IXD3235/36/37 B, C, D, E, F, and G series can quickly discharge the output capacitor (C_L) to avoid application malfunction, when CE pin set logic LOW to disable IC.

 C_L Discharge Time is proportional to the resistance (R) of the N-channel transistor located between the L_X pin and ground and the output C_L capacitance as shown below.

$$t_{DSH} = RC_L x Ln (V_{OUT(E)} / V)$$
, where

 $\label{eq:V-Output} \begin{array}{l} V \mbox{ - Output voltage after discharge} \\ V_{OUT(E)} \mbox{ - Output voltage} \\ R \mbox{ = 300 } \Omega \mbox{ (Typical value)} \end{array}$

Output Voltage Discharge Characteristics





CE/MODE Pin Function

The IXD3235/36/37 series enter the shut down mode, when a LOW logic-level signal applies to the CE/MODE pin. In the shutdown mode, IC current consumption is ~0 μ A (Typical value), with the Lx and V_{OUT} pins at high impedance state. The IC starts its operation when a HIGH logic-level signal applies to the CE/MODE pin.

Intermediate voltage, generated by external resistive divider can be used to select PWM/PFM auto or PWM only switching modes in respect with the table below.

CE/MODE VOLTAGE LEVEL	OPERATION MODE						
	IXD3235	IXD3236	IXD3537				
$0.65 \text{ V} \leq \text{V}_{\text{CE/MODE}} \leq 6.0 \text{ V}$	Synchronous Fixed PWM mode	Synchronous PWM/PFM auto switching mode	-				
$V_{IN} - 0.25 \text{ V} \leq V_{CE/MODE} \leq V_{IN}$	-	-	Synchronous PWM/PFM auto switching mode				
$0.65 \text{ V} \le \text{V}_{CE/MODE} \le \text{V}_{IN} - 1.0 \text{ V}$	-	-	Synchronous Fixed PWM mode				
$0 \text{ V} \leq \text{V}_{\text{CE/MODE}} \leq 0.25 \text{ V}$	Standby mode	Standby mode	Standby mode				

Examples of CE/MODE pin use are shown below. Please set the value of each resistor from few hundreds $k\Omega$ to few hundred M Ω . For switches, CPU open-drain I/O port and transistor can be used.

The CE/MODE pin is a CMOS input with a sink current ~ 0 μ A.

IXD3235/36 series - Examples of how to use CE/MODE pin

SW-CE	IC STATUS					
POSITION	SCHEMATIC A	SCHEMATIC B				
ON	Standby	Active				
OFF	Active	Standby				

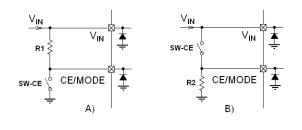
IXD3237 series - Examples of how to use CE/MODE pin

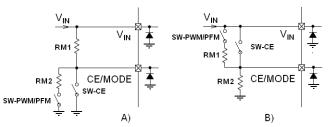
SW-CE	SW-PWM/PFM	IC STATUS			
POSITION	POSITION	SCHEMATIC A	SCHEMATIC B		
ON	v	Standby	PWM/PFM Auto		
UN	^	Standby	Switching Mode		
OFF	ON	PWM Mode	PWM Mode		
OFF	OFF	PWM/PFM Auto Switching Mode	Standby		

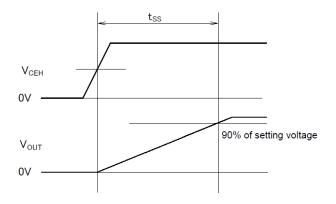
Soft Start

Soft start time is available in two options via product selection.

The soft-start time of IXD3235/36/37 series is optimized by using internal circuits and it is 1.0 ms (Typically.) for A/C/D/E series and 0.25 ms for B/F/G series. D and F series require external resistors and a capacitor to set the output voltage, so the soft-start time might vary based on value of those external components. The definition of the soft-start time is the time when the output voltage goes up to the 90% of nominal output voltage after the IC is enabled by CE "H" signal.

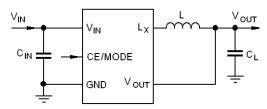




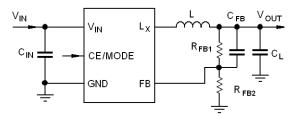


TYPICAL APPLICATION CIRCUITS

IXD3235/36/37 A, B, C, E, G Series (Fixed Output Voltage)



IXD3235/36/37 D, F Series (Adjustable Output Voltage)



EXTERNAL COMPONENTS

f _{osc}	1.2 MHz	3.0 MHz
L, µH	4.7	1.5
C _{IN} , μF	4.7	4.7
C _L , μF	10	10

Setting Output Voltage

The IXD3235/36/37 D, F Series allows set output voltage externally by two resistors R_{FB1} and R_{FB2} as sown on schematic diagram above.

Output voltage can be set starting from 0.9V. However, when input voltage (V_{IN}) is lower than the set output voltage, output voltage (V_{OUT}) cannot be higher than the input voltage.

$$\begin{split} V_{\text{OUT}} = 0.8 \times (R_{\text{FB1}} + R_{\text{FB2}}) / R_{\text{FB2}} \\ R_{\text{FB1}} + R_{\text{FB2}} < 1 \ \text{M}\Omega. \end{split}$$

The value of the phase compensation capacitor C_{FB} is calculated by the follow equation

$$f_{ZFB} = 1/(2 \times \pi \times C_{FB} \times R_{FB1}),$$

where $f_{ZFB} < 10$ kHz. For optimization, f_{ZFB} can be adjusted in the range of 1 kHz to 20 kHz depending on the inductance L and the load capacitance C_L.

Example:

When $R_{FB1} = 470 \text{ k}\Omega$ and $R_{FB2} = 150 \text{ k}$, $V_{OUT} = 0.8 \times (470 \text{ k}+150 \text{ k}) / 150 \text{ k} = 3.3 \text{ V}$

V _{out} , V	R _{FB1} , kΩ	R _{FB2} , kΩ	C _{FB} , pF	V _{OUT} , V	R _{FB1} , kΩ	R _{FB2} , kΩ	C _{FB} , pF
0.9	100	820	150	2.5	510	240	100
1.2	150	300	100	3.0	330	120	150
1.5	130	150	220	3.3	470	150	100
1.8	300	240	150	4.0	120	30	470

LAYOUT AND USE CONSIDERATIONS

- Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance. Please, pay special attention to the V_{IN} and GND wiring. Switching noise, which occurs from the GND, may cause the instability of the IC, so, position V_{IN} and V_{CL} capacitors as close to IC as possible.
- 2. Transitional voltage drops or voltage rising phenomenon could make the IC unstable if ratings are exceeded.
- 3. The IXD3235/36/37 series are designed to work with ceramic output capacitors. However, if the difference between input and output voltages is too high, a ceramic capacitor may fail to absorb the resulting high switching energy and oscillation could occur. In this case, connect an electrolytic capacitor in parallel to ceramic one to compensate for insufficient capacitance.
- 4. In PWM mode, IC generates very narrow pulses, and there is a possibility that some cycles will be skipped completely, if the difference between V_{IN} and V_{OUT} is high.
- 5. If the difference between V_{IN} and V_{OUT} is small, IC generates very wide pulses, and there is a possibility that some cycles will be skipped completely at the heavy load current.
- 6. When dropout voltage or load current is high, Current Limit may activate prematurely that will lead to IC instability. To avoid this condition, choose inductor's value to set peak current below Current Limit threshold. Calculate the peak current according to the following formula:

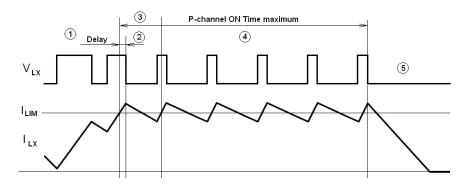
$$I_{PK} = (V_{IN} - V_{OUT}) \times D / (2 \times L \times f_{OSC}) + I_{OUT}$$
, where

L - Inductance

fosc -- Oscillation Frequency

D – Duty cycle

7. Inductor's rated current should exceed Current Limit threshold to avoid damage, which may occur until P-channel transistor turns off after Current Limiter activates (see figure below).



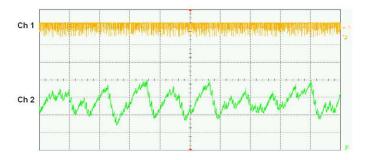
- ${\rm \textcircled{O}}$ Current flows into P-channel transistor reaches the current limit (I_{LIM}).
- [©] Current is more than I_{LIM} due the circuit's delay time from the current limit detection to the P-channel transistor OFF.
- ③ The inductor's current time rate becomes quite small.
- IC generates very narrow pulses for several milliseconds.
- ⑤ The circuit latches, stopping operation.
- 8. If V_{IN} voltage is less than 2.4 V, current limit threshold may be not reached due voltage drop caused by switching transistor's ON resistance
- 9. Latch time may become longer or latch may not work due electrical noise. To avoid this effect, the board should be laid out so that input capacitors are placed as close to the IC as possible.
- 10. Use of the IC at voltages below recommended voltage range may lead to instability.
- 11. At high temperature, output voltage may increase up to input voltage level at no load, because of the leakage current of the driver transistor.
- 12. High step-down ratio and very light load may be cause of intermittent oscillations.
- 13. In PWM/PFM automatic switching mode, IC may become unstable during transition to continuous mode. Please verify with actual components.





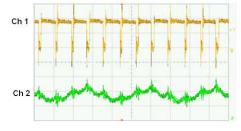
 $V_{OUT}=3.3~V,~f_{OSC}=1.2~MHz,~V_{IN}=3.7~V,~I_{OUT}=100~mA$ Ch 1 – $V_{LX}-5~V/div;$ Ch 2 – $V_{OUT}-2.0~mV/div$

External components: L = 4.7 μ H (NP4018) CIN = 4.7 μ F (ceramic) CL = 10 μ F (ceramic)



14. The IC may enter unstable operation if the combination of ambient temperature, setting voltage, oscillation frequency, and inductor's value are not adequate. If IC operates close to the maximum duty cycle, it may become unstable, even if inductor values listed below are used.

 $\begin{array}{l} {\sf V}_{{\sf OUT}}=3.3~{\sf V},~{\sf f}_{{\sf OSC}}=1.2~{\sf MHz},\\ {\sf V}_{{\sf IN}}=4.0~{\sf V},~{\sf I}_{{\sf OUT}}=150~{\sf mA}\\ {\sf Ch}~1-{\sf V}_{{\sf LX}}-2.0~{\sf V}/{\sf div};\\ {\sf Ch}~2-{\sf V}_{{\sf OUT}}-20~{\sf mV}/{\sf div}\\ {\sf External~components:}\\ {\sf L}=1.5~{\mu}{\sf H}~({\sf NP3015})\\ {\sf CIN}=4.7~{\mu}{\sf F}~({\sf ceramic})\\ {\sf CL}=10~{\mu}{\sf F}~({\sf ceramic})\\ \end{array}$

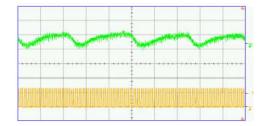


f _{osc} , MHz	V _{OUT} , V	L, µH
3.0	$0.8 \text{ V} < V_{OUT} < 4.0 \text{ V}$	1.0 – 2.2
1.2	V _{OUT} ≤2.5 V	3.3 – 6.8
1.2	V _{OUT} >2.5 V	4.7 – 6.8

If an inductor less than 4.7μ H is used at $f_{OSC} = 1.2$ MHz, or inductor less than 1.5μ H is used at $f_{OSC} = 3.0$ MHz, inductor peak current may easy reach the current limit threshold I_{LIM}. In this case, the IC may be not able to provide 600mA output current.

15. The IC may become unstable, when it goes into continuous operation mode, and difference between V_{IN} and V_{OUT} is high.

 $\begin{array}{l} {\sf V}_{{\sf OUT}}=1.8~{\sf V},\, f_{{\sf OSC}}=1.2~{\sf MHz},\\ {\sf V}_{{\sf IN}}=6.0~{\sf V},\, {\sf I}_{{\sf OUT}}=100~{\sf mA}\\ {\sf Ch}~1-{\sf V}_{{\sf OUT}}-10~{\sf mV}/{\sf div}\\ {\sf Ch}~2-{\sf V}_{{\sf LX}}-5.0~{\sf V}/{\sf div};\\ {\sf External~components:}\\ {\sf L}=4.7~{\sf \mu}{\sf H}~({\sf NP4018})\\ {\sf CIN}=4.7~{\sf \mu}{\sf F}~({\sf ceramic})\\ {\sf CL}=10~{\sf \mu}{\sf F}~({\sf ceramic})\\ \end{array}$

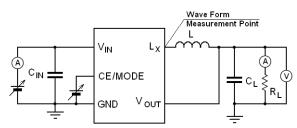


- 16. Note on mounting (WLP-5-03)
 - a) Mounting pad design should be optimized for user's conditions.
 - b) Do not use eutectics solder paste. Sn-AG-Cu solder is used for the package terminals. If eutectic solder is used, mounting reliability decreases.
 - c) When under fill agent is used to increase interfacial bonding strength, please take enough evaluation for selection. Some under fill materials and application conditions may decrease bonding reliability.
 - d) The IC has exposed surface of silicon material in the top marking face and sides, so it is weak against mechanical damages and external short circuit conditions. Please, take care of handling to avoid cracks and breaks and keep the circuit open to avoid short-circuit from the outside.
 - e) Semi-transparent resin is coated on the circuit face of the package. Please be noted that the usage under strong lights may affects device's performance.



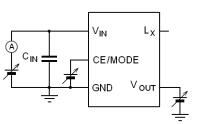
TEST CIRCUITS

Circuit ① A/B/C/E/G series

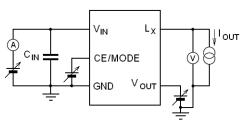


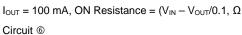
 $\begin{array}{l} \mbox{External Components} \\ \mbox{L} = 1.5 \ \mu H \ (NR3015) \ at \ 3.0 \ MHz \\ \mbox{L} = 4.7 \ \mu H \ (NR4018) \ at \ 1.2 MHz \\ \mbox{C}_{IN} = 4.7 \ \mu F \ (ceramic), \ C_L = 10 \ \mu F \ (ceramic) \end{array}$

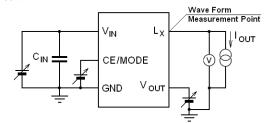
Circuit @



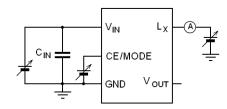
Circuit @



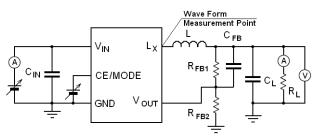




Circuit ®

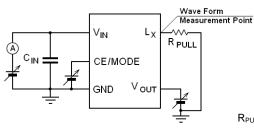


D/F series



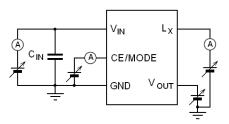
External Components L = 1.5 μ H (NR3015) at 3.0 MHz L = 4.7 μ H (NR4018) at 1.2MHz C_{IN} = 4.7 μ F (ceramic), C_L = 10 μ F (ceramic) R_{FB1} = 150 kΩ, R_{FB2} = 300 kΩ, C_{FB} = 120 pF

Circuit 3

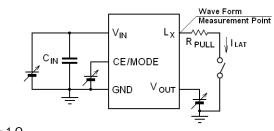


 $R_{PULL} = 200 \ \Omega$

Circuit S

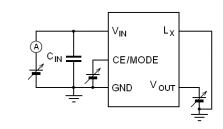


Circuit ⑦



R_{PULL} = 1 Ω Circuit ⑨

b



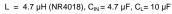


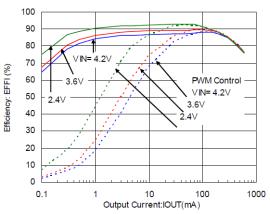
Topr = 25 $^{\circ}C$

TYPICAL PERFORMANCE CHARACTERISTICS

(1) Efficiency vs. Output Current

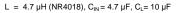
IXD3237A18C

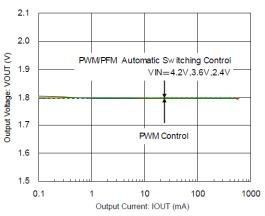




(2) Output Voltage vs. Output Current

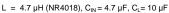
IXD3237A18C

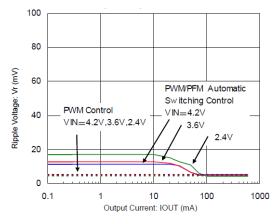


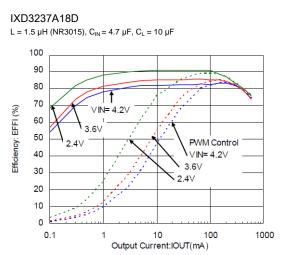


(3) Ripple Voltage vs. Output Current

IXD3237A18C

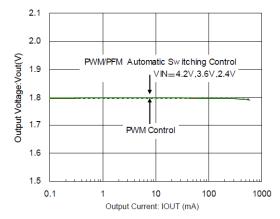






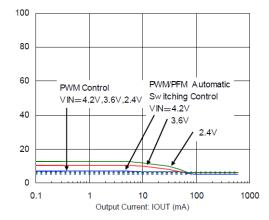
IXD3237A18D

L = 1.5 μH (NR3015), C_{IN} = 4.7 $\mu F,$ C_L = 10 μF



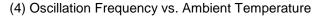
IXD3237A18D

L = 1.5 μH (NR3015), C_{IN} = 4.7 $\mu F,$ C_L = 10 μF

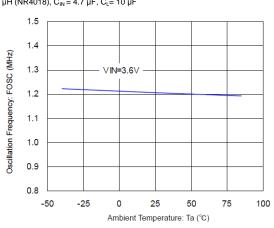


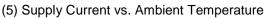


TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

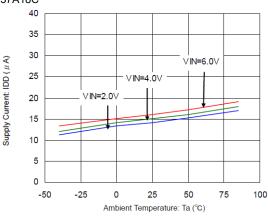


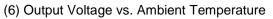




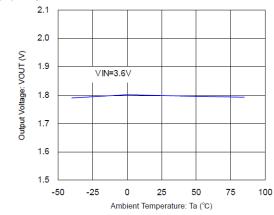


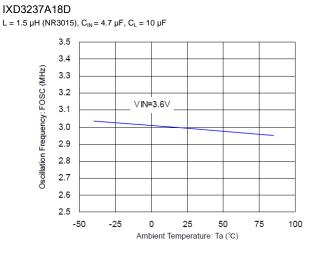




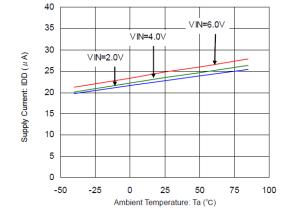


IXD3237A18D

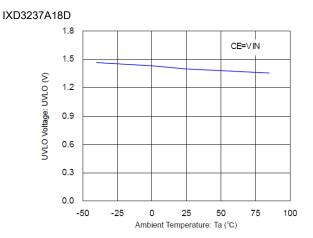




IXD3237A18D



(7) UVLO Voltage vs. Ambient Temperature

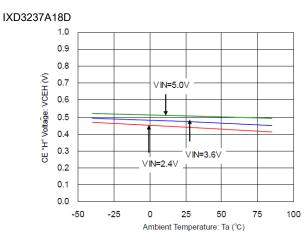




TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(8) CE "H" Voltage vs. Ambient Temparature

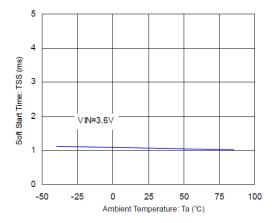
(9) CE "L" Voltage vs. Ambient Temperature

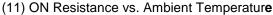


(10) Soft Start Time vs. Ambient Temperature

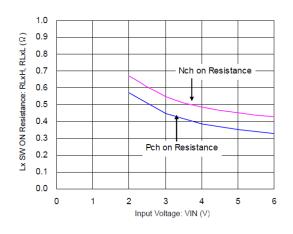
IXD3237A18C

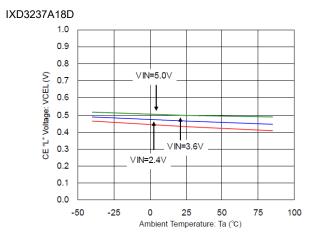
L = 4.7 μ H (NR4018), C_{IN} = 4.7 μ F, C_L= 10 μ F



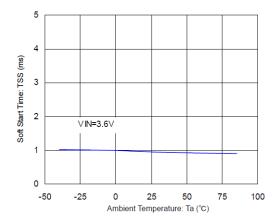


IXD3237A18D





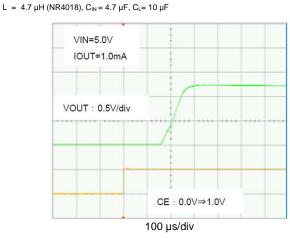
IXD3237A18D L = 1.5 μ H (NR3015), C_{IN} = 4.7 μ F, C_L = 10 μ F





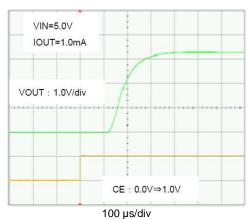
(12) IXD3235/36/37 B version Start Wave Form

IXD3237B12C



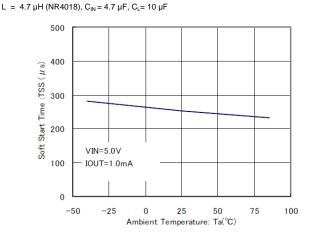
IXD3237B33D

L = 1.5 μH (NR3015), C_{IN} = 4.7 $\mu F,$ C_L = 10 μF

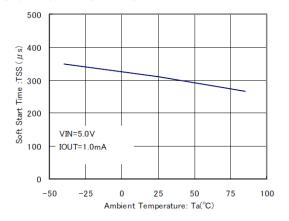


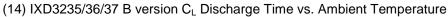
(13) IXD3235/36/37 B version Soft Start Time vs. Ambient Temperature

IXD3237B12C

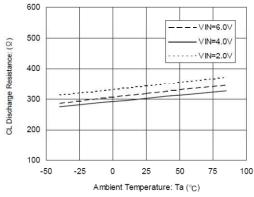


IXD3237B33D L = 1.5 μ H (NR3015), C_{IN} = 4.7 μ F, C_L = 10 μ F





IXD3237B33D



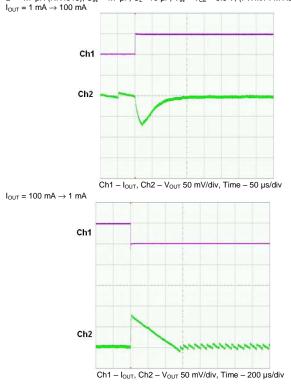


TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(15) Load Transient Response

IXD3237A18C

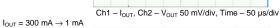
L = 4.7 μ H (NR4018), C_{IN} = 4.7 μ F, C_L= 10 μ F, V_{IN} = V_{CE} = 3.6 V, (PWM/PFM Auto Switching mode) $I_{OUT} = 1 \text{ mA} \rightarrow 300 \text{ mA}$





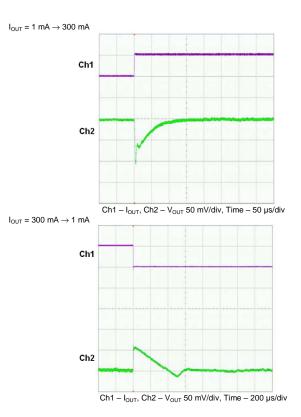
Ch1

Ch2

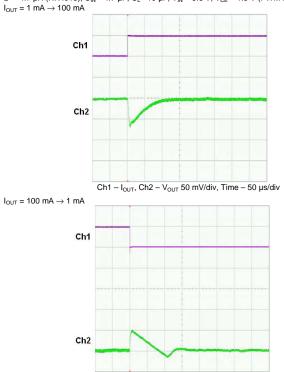




 $Ch1-I_{OUT},\,Ch2-V_{OUT}$ 50 mV/div, Time – 200 $\mu s/div$



IXD3237A18C L = 4.7 μH (NR4018), C _{IN} = 4.7 $\mu F,~C_L =$ 10 $\mu F,~V_{IN} =$ 3.6 V, V $_{CE}$ = 1.8 V (PWM mode)



Ch1 - I_{OUT}, Ch2 - V_{OUT} 50 mV/div, Time - 200 µs/div



TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(15) Load Transient Response (Continued)

IXD3237A18D

L = 1.5 μ H (NR3015), C_{IN} = 4.7 μ F, C_L= 10 μ F, V_{IN} = V_{CE} = 3.6 V, (PWM/PFM Auto Switching mode)



Ch1 – I_{OUT}, Ch2 – V_{OUT} 50 mV/div, Time – 200 μ s/div

Ch2

Ch2

Ch1 – $I_{\text{OUT}},$ Ch2 – V_{OUT} 50 mV/div, Time – 200 $\mu s/div$



ORDERING INFORMATION

IXD3235023456-7 IXD3236023456-7 IXD3237023456-7

DESIGNATOR	DESCRIPTION	SYMBOL	DESCRIPTION
		A B C	
0	Type of DC/DC Controller	E	Refer to Product Classification
		G D	
		F	
23	Fixed Output Voltage, V	08 - 40	 ② - integer part, ③ - decimal part, i.e. V_{OUT} = 2.8 V - ② = 2, ③ = 8 V_{OUT} = 2.85 V - ② = 2, ③ = L 0.05 V increments: 0.05 = A, 0.15 = B, 0.25 = C. 0.35 = D, 0.45 = E, 0.55 = F, 0.65 = H, 0.75 = K, 0.85 = L, 0.95 = M
	Reference Voltage	08	Reference Voltage (Fixed) 0.8 V - 2 = 0, 3 = 8
(4)	Oscillation Frequency	C	1.2 MHz 3.0 MHz
\$6-7*	Packages (Order Limit)	MR	SOT-25 (3000/reel)
		MR-G	SOT-25 (3000/reel)
		ER	USP-6C (3000/reel)
		ER-G	USP-6C (3000/reel)
		4R-G	USP-6EL (3000/reel)
		0R-G	WLP-5-03 (3000/reel)

NOTE:

1) The "-G" suffix denotes halogen and antimony free, as well as being fully RoHS compliant.

- 2) SOT-25, USP-6EL package are available for the A/B/C series only.
- 3) WLP-5-03 package is available for the A/B series only.

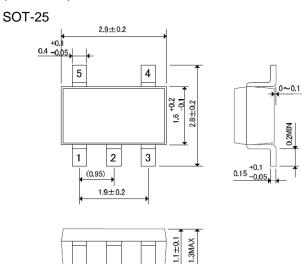
PRODUCT CLASSIFICATION

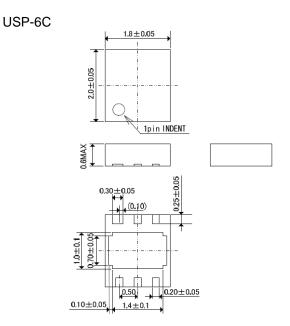
Туре	vo	DUT	V	IN	C∟ Auto discharge	Soft	Start
	Fixed	Adjustable	≥ 1.8 V	≥ 2 V		High Speed	Low Speed
Α	Yes	No	No	Yes	No	No	Yes
В	Yes	No	No	Yes	Yes	Yes	No
С	Yes	No	No	Yes	Yes	No	Yes
D	No	Yes	Yes	No	Yes	No	Yes
E	Yes	No	Yes	No	Yes	No	Yes
F	No	Yes	Yes	No	Yes	Yes	No
G	Yes	No	Yes	No	Yes	Yes	No



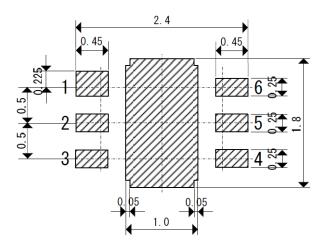
PACKAGE DRAWING AND DIMENSIONS

(Units: mm)

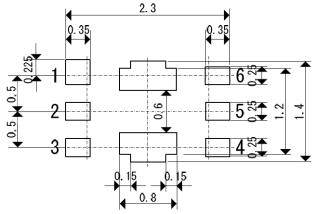




USP-6C Reference Pattern Layout



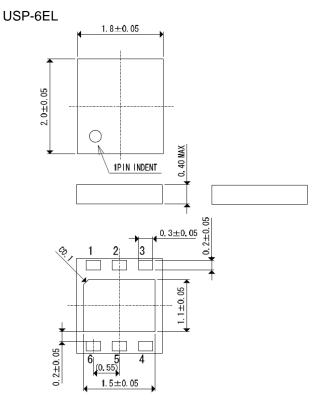
USP-6C Reference Metal Mask Design

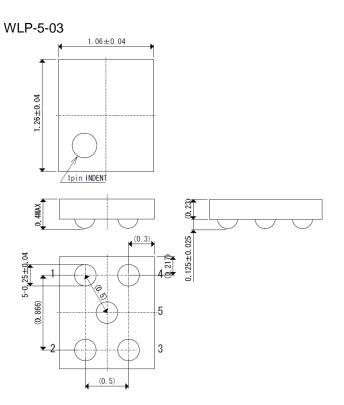




PACKAGE DRAWING AND DIMENSIONS (CONTINUED)

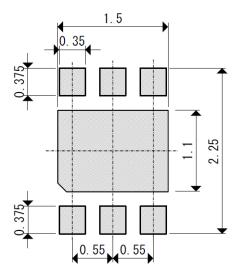
(Units: mm)



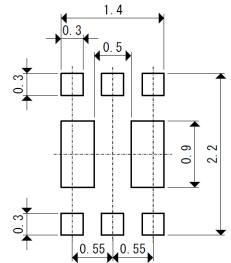


NOTE: A part of the pin may appear from the side of the package because of its structure, but reliability of the package and strength will be not below the standard.

USP-6EL Reference Pattern Layout



USP-6EL Reference Metal Mask Design



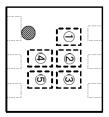
IXYSPOWERICs

MARKING

SOT-25

5 4 (1) (2) (3) (4) (6) 1 2 3 (TOP VIEW)

USP-6C/USP-6EL



(TOP VIEW)

WLP-5-03

1		2	
\bigcirc	\rangle	Θ	
		$\overline{\mathbf{N}}$	
	୦	ω	
4	5	3	

(TOP VIEW)

① Represents product series

PRODUCT SERIES	MARK				
PRODUCT SERIES	IXD3235	IXD3236	IXD3237		
A	4	5	6		
В	С	D	E		
С	K	L	М		
D	K	L	М		
E	4	5	6		
F	2	7	В		
G	С	D	E		

^② Represents integer number of the output voltage and oscillation frequency

A/B/C/F series

V V	MARK			
V _{out} , V	f _{osc} = 1.2 MHz	$f_{OSC} = 3.0 \text{ MHz}$		
0.x	A	F		
1.x	В	Н		
2.x	С	К		
3.x	D	L		
4.x	E	М		

E/G/D Series

V V	MARK			
V _{out} , V	f _{osc} = 1.2 MHz	$f_{OSC} = 3.0 \text{ MHz}$		
0.x	A	F		
1.x	В	Н		
2.x	С	К		
3.x	D	L		
4.x	E	М		

③ Represents decimal value of the output voltage

V _{OUT} , V	MARK	V _{OUT} , V	MARK
x.00	0	x.05	A
x.10	1	x.15	В
x.20	2	x.25	С
x.30	3	x.35	D
x.40	4	x.45	E
x.50	5	x.55	F
x/60	6	x.65	Н
x.70	7	x.75	K
x.80	8	x.85	L
x.90	9	X,95	М

④⑤ represents production lot number

01~09、0A~0Z、11~9Z、A1~A9、AA~AZ、B1~ZZ in order (G, I, J, O, Q, and W excluded)

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